

USRP testbed for spectrum sensing of OFDM signals

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2012-05-29

Outline

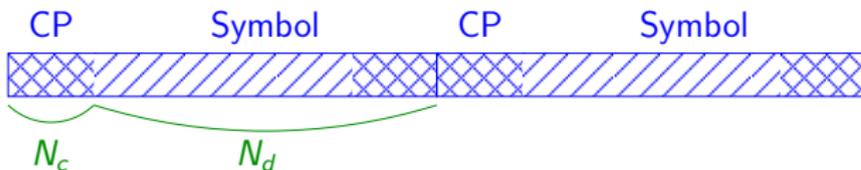
- 1 Introduction
- 2 Spectrum sensing
- 3 Measurements
- 4 Conclusions

Background

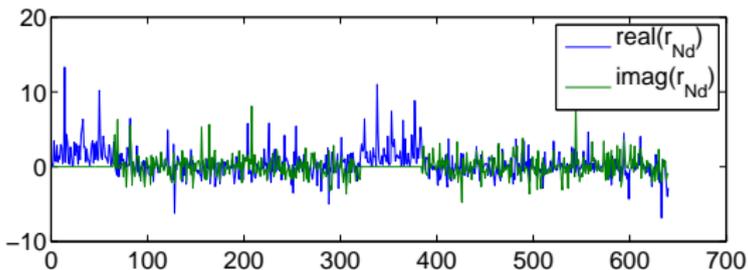
- Cognitive radio
 - opportunistic use of licensed spectrum by secondary user
 - autonomous units with adaptable radio-system parameters
 - requires ability to detect primary user activity
- Application: secondary use of TV frequencies
 - TV frequencies often underutilized
 - IEEE 802.22: rural broadband access
 - detection of primary user: spectrum sensing or (national) database
- Spectrum sensing
 - detection of primary user in licensed spectrum
 - does not require legacy channel to database
 - finer detection granularity (in time and space)
- Focus of work
 - spectrum sensing of digital TV (OFDM) signals
 - practical evaluation of sensing algorithms
 - single secondary user, idle while sensing

Feature-based signal detection

- Primary user uses OFDM signal
- FFT size: N_d , cyclic prefix: N_c



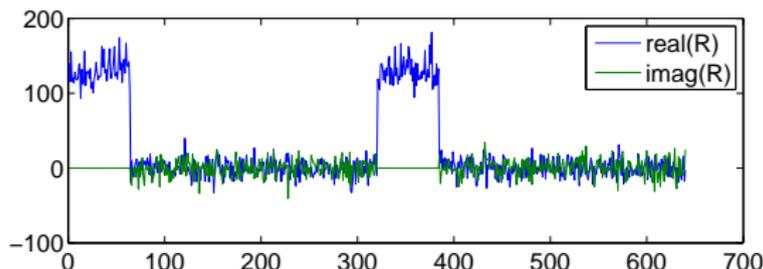
- Received signal: $x(n)$
- Define auto-correlation at distance N_d : $r_{N_d}(n) = x(n)x^*(n + N_d)$



Feature-based signal detection (cont)

- Noise suppression by averaging over K symbols:

- $R(n) = \sum_{k=0}^{K-1} r_{N_d}(n + k(N_d + N_c)), n = 0, \dots, N_d + N_c - 1$



- General description of sensing algorithm

- Compute metric M based on $x(n)$, $r_{N_d}(n)$ and/or $R(n)$
- Primary user detected if $M > t$
- t threshold calibrated such that $P(M > t) = P_{FA}$ (false alarm probability) when primary user not present

Algorithms

- Averaging

$$M = \frac{1}{\sum_{n=0}^{K(N_d+N_c)-1} |x(n)|^2} \left| \sum_{n=0}^{N_d+N_c-1} R(n) \right|$$

- Sliding window [802.22]

$$M = \frac{1}{\sum_{n=0}^{K(N_d+N_c)-1} |x(n)|^2} \max_{\tau} \left| \sum_{n=\tau}^{\tau+N_c-1} R(n) \right|$$

- Generalized likelihood ratio test-based

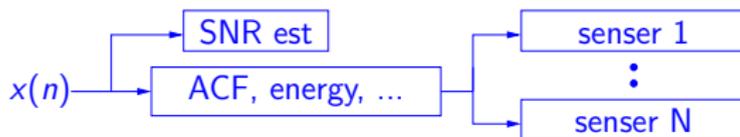
$$M = \max_{\tau} \frac{\sum_{i=0}^{N_c+N_d-1} |R(n)|^2}{\sum_{k \in S_{\tau}} \left| R(k) - \frac{1}{N_c} \sum_{i \in S_{\tau}} R(i) \right|^2 + \sum_{j \notin S_{\tau}} |R(j)|^2}$$

- Energy

$$M = \sum_{n=0}^{K(N_d+N_c)-1} |x(n)|^2$$

USRP implementation

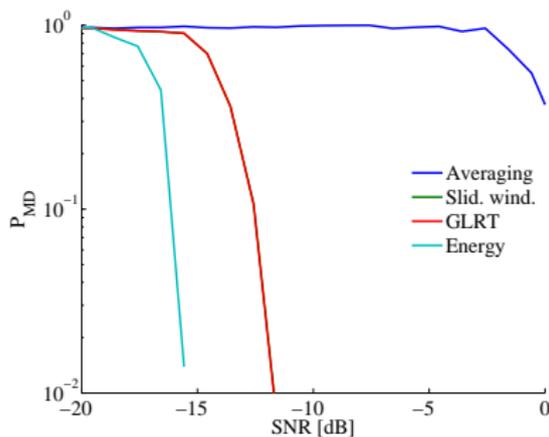
- Measurement setup
 - 1 USRP as primary user, 1 USRP as secondary user
 - secondary user is only sensing the spectrum
 - USRP 1 with RFX2400 daughterboards
 - Measurements done in university basement (weak WLAN signals present)
 - Antenna distance: ca 10 meters
- Spectrum sensor data path



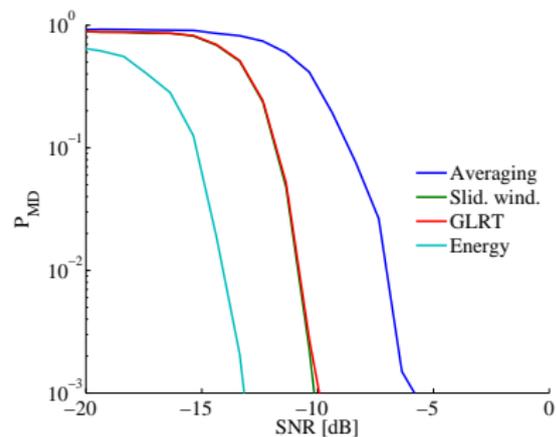
- Receiver SNR estimation
 - SNR range: -30, .. -10 dB: SNR estimation hard
 - SNR estimation algorithm
 - 1 Measure received P_{noise} with transmitter off
 - 2 Measure received P_{fs} with transmitter at full power
 - 3 Compute $SNR_{0dBfs} = 10 \log_{10} \frac{P_{fs} - P_{noise}}{P_{noise}}$
 - 4 Determine SNR at A dBfs: $SNR = SNR_{0dBfs} + A$

Measurement results

- Signal bandwidth: 6.4 MHz
- Calibration for $P_{FA} = 0.05$



- FFT: $N_d = 2048$
- Cyclic prefix: $N_c = 64$
- Sensing time: 16.9 ms ($K = 64$)

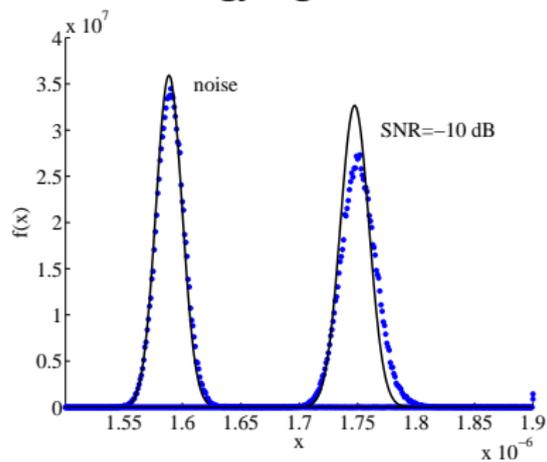


- FFT: $N_d = 256$
- Cyclic prefix: $N_c = 64$
- Sensing time: 2.56 ms ($K = 64$)

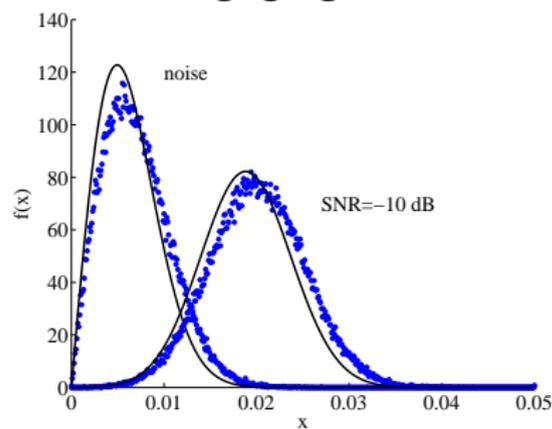
Metric distribution

- Theoretical distribution of metrics determined for energy and averaging algorithms
- Can be used to set calculate threshold theoretically

Energy algorithm



Averaging algorithm



Conclusions

- General observations
 - Sliding window and GLRT-based detectors very similar in performance
 - Averaging detector inferior
 - Energy detector superior despite being sensitive to noise estimation
 - WRAN detector similar to GLRT detector
- Future work
 - Outdoor measurements
 - Performance in presence of interference
 - Measurements with larger FFTs
 - Robustness to noise uncertainty