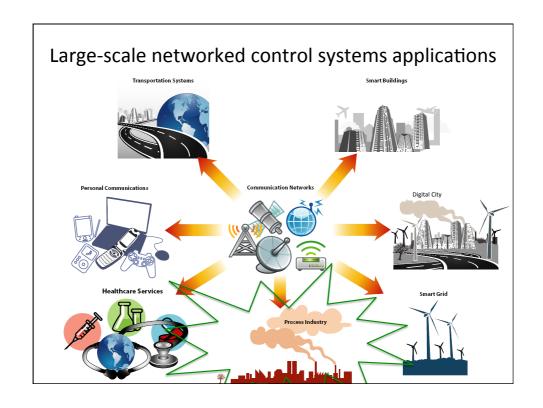


Wireless Control: Medium Access and Event-based Control

Karl Henrik Johansson ACCESS Linnaeus Center Royal Institute of Technology, Sweden



Academy of Mathematics and Systems Science, CAS, Beijing, Jul 4, 2012



Outline

- Introduction
- Industrial applications of wireless control
- Medium access
- Event-based control
- Conclusions

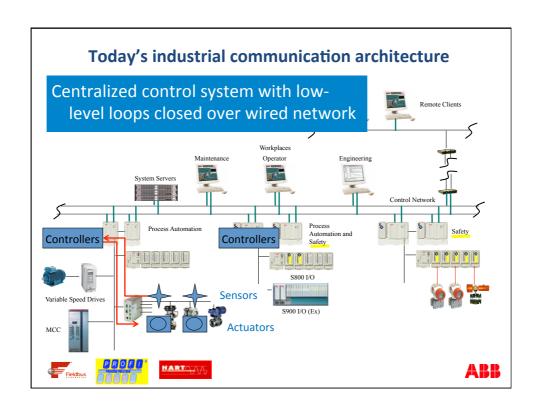
Communication in process control Dominant communication technology 3-15 psi Pneumatic Analog Analog Smart Fieldbuses Wireless HART ISA100 ZigBee [ISA100]

Wireless sensor systems benefit from

- Lower installation and maintenance costs
- · Increased sensing capabilities and flexibility

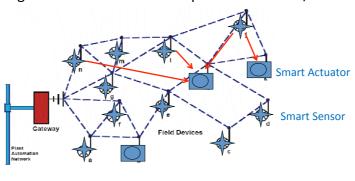
Major consequences for control system architectures

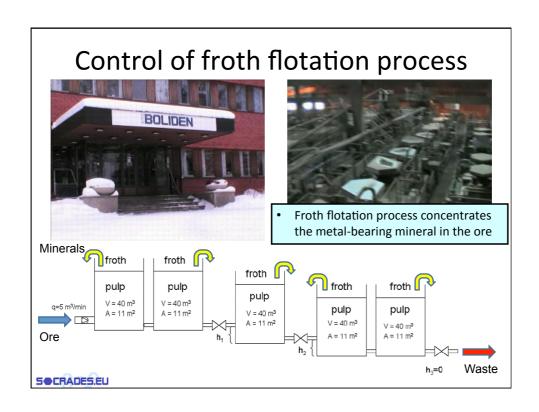


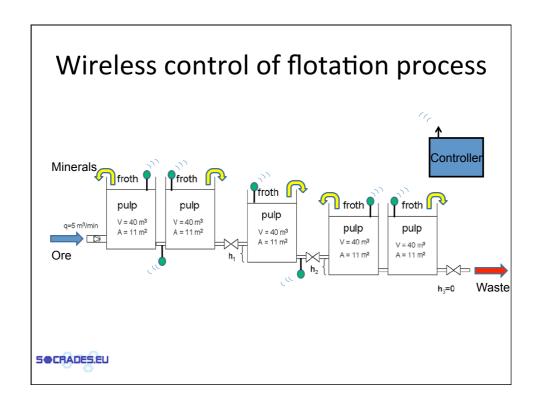


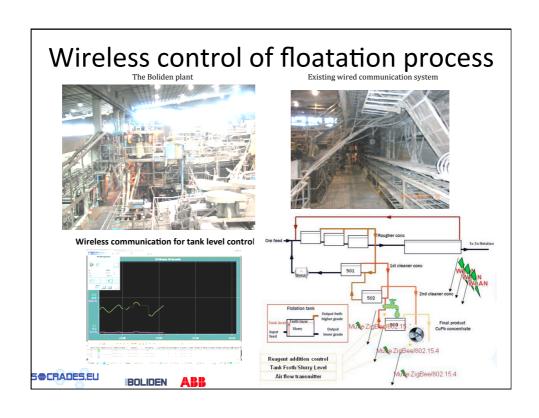
Towards wireless sensor and actuator network architecture

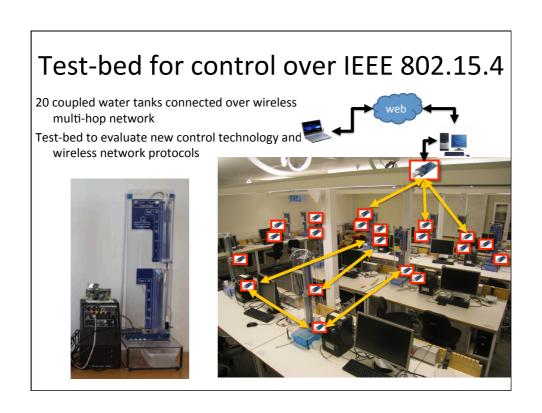
- Local control loops closed over wireless multi-hop network
- Potential for a dramatic change:
 - From fixed hierarchical centralized system to flexible distributed
 - Move intelligence from dedicated computers to sensors/actuators





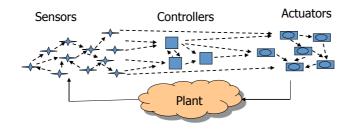






Wireless control system

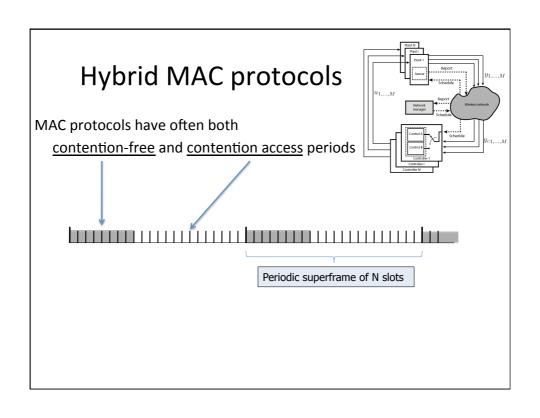
How share common network resources while maintaining guaranteed closed-loop performance?

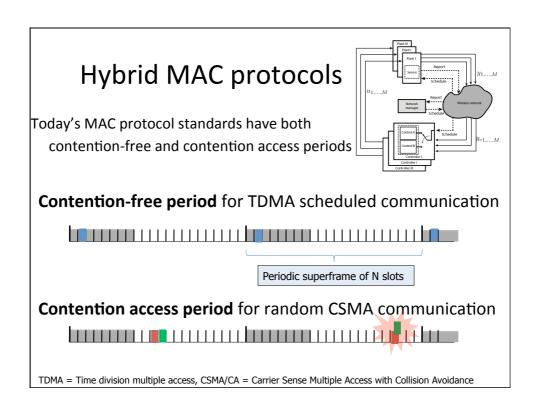


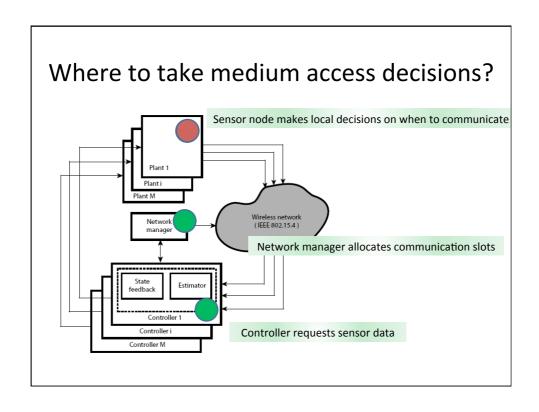
- How handle network imperfections: resource constraints, loss, conflicts, delays, outages?
- How move intelligence from a few central units to many distributed devices?

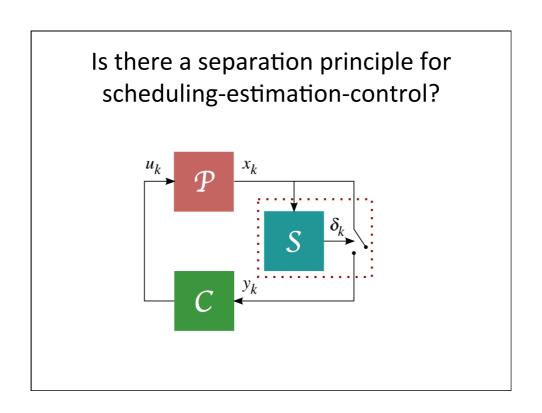
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Stochastic control formulation

Plant:

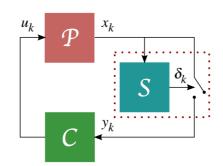
$$x_{k+1} = Ax_k + Bu_k + w_k$$

Scheduler:

$$\begin{split} & \delta_k = f_k(\mathbb{I}_k^{\mathbb{S}}) \in \{0,1\} \\ & \mathbb{I}_k^{\mathbb{S}} = \left[\{x\}_0^k, \{y\}_0^{k-1}, \{\delta\}_0^{k-1}, \{u\}_0^{k-1} \right] \end{split}$$

Controller

$$\begin{aligned} u_k &= g_k(\mathbb{I}_k^{\mathbb{C}}) \\ \mathbb{I}_k^{\mathbb{C}} &= \left[\{y\}_0^k, \{\delta\}_0^k, \{u\}_0^{k-1} \right] \end{aligned}$$

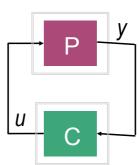


Cost criterion:

$$J(f,g) = \mathbf{E}[x_N^T Q_0 x_N + \sum_{s=0}^{N-1} (x_s^T Q_1 x_s + u_s^T Q_2 u_s)]$$

Certainty equivalence revisited

Definition Certainty equivalence holds if the closed-loop optimal controller has the same form as the deterministic optimal controller with x_k replaced by the estimate $\hat{x}_{k|k} = \mathrm{E}[x_k|\mathbb{I}_k^{\mathbb{C}}]$.



Theorem[Bar-Shalom–Tse] Certainty equivalence holds if and only if $E[(x_k - E[x_k|I_k^c])^2|I_k^c]$ is not a function of past controls $\{u\}_0^{k-1}$ (no dual effect).

Feldbaum, 1965; Åström, 1970; Bar-Shalom and Tse, 1974

State-based scheduler

Plant:

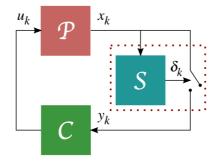
$$x_{k+1} = Ax_k + Bu_k + w_k$$

Scheduler:

$$\begin{split} & \delta_k = f_k(\mathbb{I}_k^{\mathbb{S}}) \in \{0, 1\} \\ & \mathbb{I}_k^{\mathbb{S}} = \left[\{x\}_0^k, \{y\}_0^{k-1}, \{\delta\}_0^{k-1}, \{u\}_0^{k-1} \right] \end{split}$$

Controller:

$$\begin{aligned} u_k &= g_k(\mathbb{I}_k^{\mathbb{C}}) \\ \mathbb{I}_k^{\mathbb{C}} &= \left[\{y\}_0^k, \{\delta\}_0^k, \{u\}_0^{k-1} \right] \end{aligned}$$



Corollary The control u_k for the optimal closed-loop system has a dual effect.

The separation principle does not hold for the optimal closed-loop system, so the design of the scheduler, estimator, and controller is coupled

Ramesh, Sandberg, Bao, J, 2009, 2010

Conditions for Certainty Equivalence

Corollary: The optimal controller for the system $\{P,S(f),C(g)\}$, with respect to the cost J is certainty equivalent if and only if the scheduling decisions are not a function of the applied controls.



Ramesh, Sandberg, Bao, J, 2011

20

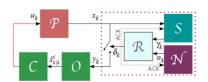
Observer-Scheduler for Certainty Equivalence

A symmetric scheduling policy results in separation between the estimator and the scheduler, as well as an optimal certainty equivalent

Observer: $\hat{x}_{k|k} = \begin{cases} x_k & \delta_k = 1 \\ A^{k-\tau_k} x_{\tau_k} + \sum_{s=1}^{k-\tau_k} A^{s-1} B u_{k-s} & \delta_k = 0 \\ + \mathbb{E}[\sum_{s=1}^{k-\tau_k} A^{s-1} w_{k-s} | \hat{f}_k, ..., \hat{f}_{\tau_k+1} = 0] & \delta_k = 0 \end{cases}$

• Symmetric scheduler:

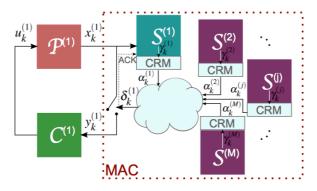
$$\gamma_k = f_{|\cdot|}(\sum_{s=1}^{k-\tau_{k-1}} A^{s-1} w_{k-s}) \qquad \forall k \; , \\ \text{where, } f_{|\cdot|}(r) = f_{|\cdot|}(-r)$$



• Certainty equivalence achieved at the cost of optimality Ramesh, Sandberg, Bao, J, 2011

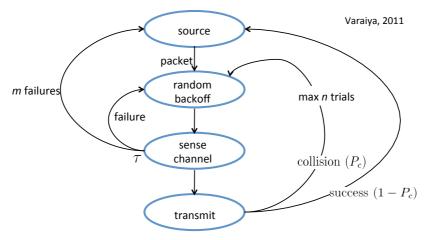
Extensions to multiple loops and contention resolution mechanisms

- \bullet Hard problem because of the correlation between the plants imposed by the MAC
- Closed-loop analysis can still be done for a class of event-based schedulers and simple MAC's
- General problem with event-based schedulers and realistic MAC (e.g., CSMA/CA) is open



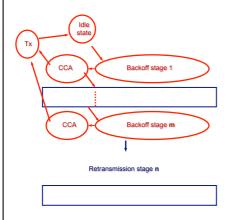
Ramesh et al, 2011

How to model CSMA/CA MAC?



- · Every device executes this protocol
- Assume all carrier sense events are independent [Bianchi, 2000]

CSMA/CA mechanism of a node in an IEEE 802.15.4 wireless network

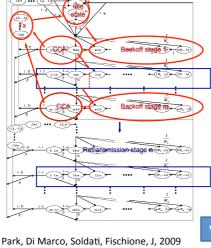


- A transmitting node delays for a random number of backoff periods in [0, 2^m.-1], where m_a is the initial backoff exponent.
- If two consecutive clear channel assessments (CCA) are idle, the node starts the transmission and waits for an ACK
- If the channel is busy, the procedure is repeated increasing the backoff windows until a maximum backoff exponent m_b.
- After a maximum number of backoffs m the packet is discarded.
- In case of collision the procedure is restarted and repeated until a retry limit n

Park, Di Marco, Soldati, Fischione, J, 2009

Cf., Bianchi, 2000; Pollin et al., 2006

Markov chain model of CSMA/CA



- Markov state (s,c,r)
 - s: backoff stage
 - c: state of backoff counter
 - r: state of retransmission counter
- Model parameters
 - q_0 : traffic condition (q_0 =0 saturated)
 - − m₀, m, m_b, n: MAC parameters
- Computed characteristics
 - α: busy channel probability during CCA1
 - 6: busy channel probability during CCA2
 - P_c: collision probability

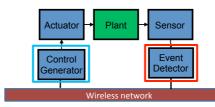
Validated in simulation and experiment

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When to transmit?

- Medium access control-like mechanism at sensor
 - E.g., threshold crossing, adaptive sampling



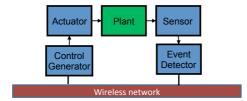
How to control?

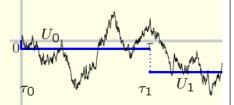
- Execute control law over fixed control alphabet
 - E.g., piecewise constant controls, impulse control

Åström, 2007, Rabi and J., WICON, 2008

Example: Fixed threshold with impulse control Event-detector implemented as fixedlevel threshold at sensor Event Detector Control Event-based impulse control better than periodic impulse control Event-Based Control 200 200 100 100 -100 -100-200 L 5 10 15 20 Åström & Bernhardsson, *IFAC*, 1999

Event-based ZoH control with adaptive sampling





How choose $\{U_i\}$ and $\{\tau_i\}$ to minimize $V=\frac{1}{T}E\int_0^T x^2(t)dt$.

Rabi, J, Johansson, 2008

Controlled Brownian motion with one sampling event

$$dx_t = u_t dt + dB_t$$

$$\min_{U_0, U_1, \tau} J = \min_{U_0, U_1, \tau} \mathbf{E} \int_0^T x_s^2 ds$$



$$= \min_{U_0, U_1, \tau} \left[\mathbf{E} \int_0^\tau x_s^2 ds + \mathbf{E} \int_\tau^T x_s^2 ds \right]$$

A joint optimal control and optimal stopping problem

Rabi, J, Johansson, 2008

$$\begin{split} dx_t &= u_t dt + dB_t \\ \min_{U_0, U_1, \tau} J &= \min_{U_0, U_1, \tau} \mathbf{E} \int_0^T x_s^2 ds \end{split}$$



If τ chosen deterministically (not depending on x_t) and $x_0 = 0$:

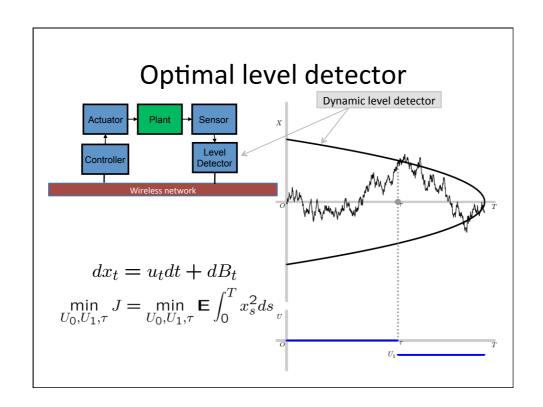
$$U_0^* = 0$$
 $U_1^* = -\frac{3x_{T/2}}{T}$ $\tau^* = T/2$

If τ is event-driven(depending on x_t) and $x_0 = 0$:

$$U_0^* = 0$$
 $U_1^* = -\frac{3x_{\tau^*}}{2(T - \tau^*)}$

$$\tau^* = \inf\{t : x_t^2 \ge \sqrt{3}(T - t)\}$$

Envelope defines optimal level detector



Policy iteration

For $x_0 \neq 0$ we have in general the cost function

$$J_N\left(x_0, \{U_0, U_1\}, \tau\right) \stackrel{\Delta}{=} \alpha\left(x_0, T\right) - \mathbb{E}\left[\beta\left(x_0, U_0, \tau, T\right)\right],$$

where

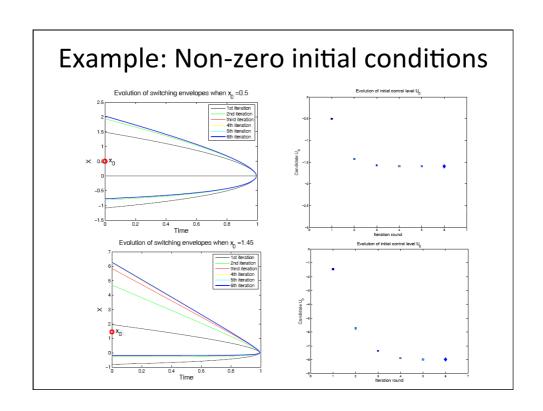
$$\begin{split} \alpha\left(x_0, U_0, T\right) &= \int_0^T \mathbb{E}\left[\Phi_{U_0}^2(s, 0, x_0)\right] ds \\ \beta\left(x_0, U_0, \tau, T\right) &= \int_\tau^T \mathbb{E}\left[\Phi_{U_0}^2(s, \tau, x_\tau) - \Phi_{U_1^*(x_\tau, \tau, T)}^2(s, \tau, x_\tau)\right] \end{split}$$

and $\Phi_U(t_2,t_1,x)$ is the solution of the system with constant control

Necessary condition for optimality

$$\begin{cases} \tau^* \left(x_0 \right) &= \operatorname{ess \, sup \, } \mathbb{E} \left[\beta \left(x_0, U_0^* \left(x_0 \right), \tau, T \right) \right], \\ U_0^* \left(x_0 \right) &= \inf_{U} \left\{ \alpha \left(x_0, U, T \right) - \mathbb{E} \left[\beta \left(x_0, U, \tau^* \left(x_0 \right), T \right) \right] \right\}. \end{cases}$$

suggests iterative search algorithm. Computationally intensive.



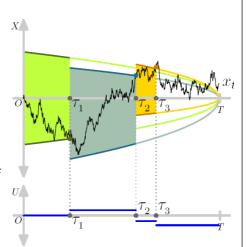
Multiple samples

Extension to N>1 samples

$$J_{N}\left(x_{0},\mathcal{U},\left\{ \tau\right\} _{i=1}^{N}\right)=\mathbb{E}\left[\left.\int_{0}^{T}x_{s}^{2}ds\right|x_{0}\right]$$

through nested single sample problems

Extensions to variable budget sampling, allowing number of samples to depend on x.

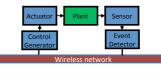


Event-based impulse control

Plant

$$dx_t = dW_t + u_t dt, \ x(0) = x_0,$$

Sampling events
$$\mathcal{T} = \left\{ au_0, au_1, au_2, \ldots
ight\},$$



Impulse control $u_t = \sum_{n=0}^{\infty} x_{\tau_n} \delta\left(\tau_n\right)$

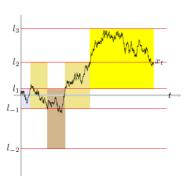
 $\text{Average sampling rate} \quad R_{\tau} = \limsup_{M \to \infty} \frac{1}{M} \mathbb{E} \left[\int_{0}^{M} \sum_{n=0}^{\infty} \mathbf{1}_{\{\tau_{n} \leq M\}} \delta\left(s - \tau_{n}\right) ds \right]$

Average cost $J = \limsup_{M \to \infty} \frac{1}{M} \mathbb{E} \left[\int_0^M x_s^2 ds \right]$

Level-triggered control

Ordered set of levels $\mathcal{L}=\{\ldots,l_{-2},l_{-1},l_0,l_1,l_2,\ldots\}$ $l_0=0$ Multiple levels needed because we allow packet loss

Sampling instances $\tau = \inf \left\{ \tau \middle| \tau > \tau_i, x_\tau \in \mathcal{L}, x_\tau \notin x_{\tau_i} \right\}$



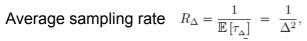
Level-triggered control

For Brownian motion, equidistant sampling is optimal

$$\mathcal{L}^* = \{k\Delta | k \in \mathbb{Z}\}$$

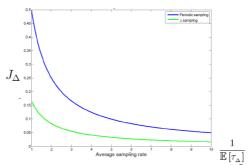
First exit time

$$\tau_{\Delta} = \inf \left\{ \tau \middle| \tau \ge 0, x_{\tau} \notin (\xi - \Delta, \xi + \Delta), x_{0} = \xi \right\}$$



Average cost
$$J_{\Delta} = \frac{\mathbb{E}\left[\int_{0}^{\tau_{\Delta}} x_{s}^{2} ds\right]}{\mathbb{E}\left[\tau_{\Delta}\right]} = \frac{\Delta^{2}}{6}.$$

Comparison between time- and event-based control



 $T=\Delta^2$ gives equal average sampling rate for periodic control and event-based control

Event-based impulse control is three times better than periodic

Åström & Bernhardsson, 1999

What about the influence of communication losses? Is event-based sampling still better?

Influence of i.i.d. packet loss

Times when packets are successfully received $\rho_i \in \{\tau_0 = 0, \tau_1, \tau_2, \ldots\}$,

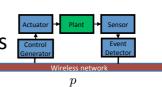
$$\{\rho_0 = 0, \rho_1, \rho_2, \ldots\}$$
. $\rho_i \geq \tau_i$,

Average rate of packet reception

$$R_{\rho} = \limsup_{M \rightarrow \infty} \frac{1}{M} \mathbb{E} \left[\int_{0}^{M} \sum_{n=0}^{\infty} \mathbf{1}_{\{\rho_{n} \leq M\}} \delta \left(s - \rho_{n}\right) ds \right] = p \cdot R_{\tau}$$

Define the times between successful packet receptions $P_{(p,\Delta)}$

Event-based control with losses



Theorem

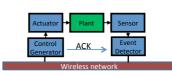
If packet losses are i.i.d. with probability p, then level-triggered sampling gives

$$J_p = \frac{\Delta^2 \left(5p + 1\right)}{6 \left(1 - p\right)}$$

Event-based control better than periodic control if loss probability

Rabi and J, 2009

Communication acknowledgements



If controller perfectly acknowledges packets to sensor, event detector can adjust its sampling strategy

Let
$$\Delta\left(l\right)=\sqrt{l+1}\Delta_{0}$$

where $l \ge 0$ number of samples lost since last successfully transmitted packet

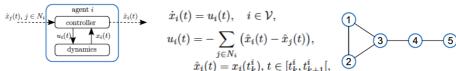
Gives that $\mathbb{E}\left[au_{i+1}^{\dagger}- au_{i}^{\dagger}
ight]$ becomes independent of i.

Better performance than fixed $\Delta(l)$ for same sampling rate:

$$J_p^{\uparrow} = \frac{\Delta^2 (1+p)}{6 (1-p)} \le \frac{\Delta^2 (1+5p)}{6 (1-p)} = J_p.$$

Rabi and J, 2009

Event-based Control for Multi-agent Systems



Trigger condition for state broadcasting: $t_{k+1}^i = \inf\{t > t_k^i: \ f_i(t,e_i(t)) > 0\}$ $e_i(t) = \hat{x}_i(t) - x_i(t).$ Theorem Suppose

$$f_i(t,e_i(t)) = |e_i(t)| - (c_0 + c_1 e^{-\alpha t}), \quad c_0,c_1 \ge 0, c_0 + c_1 > 0$$

and $0 < \alpha < \lambda_2(\mathcal{G})$. Then, the closed-loop system is non-Zeno and x(t) converges to a ball around $(1/N)\mathbf{1}\mathbf{1}^Tx(0)$ with radius $\|L\|\sqrt{N}c_0/\lambda_2(\mathcal{G})$.

Extensions to double-integrator and time-delay systems

[Seyboth et al, 2011; Dimarogonas et al, 2012]

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Conclusions

- Wireless control is an enabling technology in many emerging industrial applications
- Fundamental challenges related to
 - time-driven, synchronous, sampled data control theory, vs
 - event-driven, asynchronous, ad hoc wireless networking
- · Integrated modeling for medium access and control
- Event-based control provides a natural principle for large-scale wireless control systems

http://www.ee.kth.se/~kallej

